

# MINIMAL REALIZATIONS OF THE INVERSE OF A POLYNOMIAL MATRIX USING FINITE AND INFINITE JORDAN PAIRS

G.F. FRAGULIS†

†Department of Mathematics, Faculty of Sciences, Aristotle University of Thessaloniki, Thessaloniki, 54006, Greece.

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## Abstract

It is rather obvious that the connections between control theory and linear algebra are very strong. Several formulas and notions, as well as, known techniques from matrix theory and theory of operators are used efficiently in control theory. The important treatise of [4] gives a nice example of how matrix theory can be applied to the analysis and solution—finding of several difficult problems in control theory. On the other hand Gohberg and other researchers [5] presented their work on operator polynomial and general operator-valued functions, and pointed out the striking similarities among them and formulas and notions in control theory making the observation that "... from the systems theory point of view, we study here systems for which the transfer function matrix is the inverse of a polynomial matrix" [5, page 7]. In this paper we present a simple method which uses the notions of finite and infinite Jordan pairs from the operator theory in such a way to find a minimal realization of the inverse of a polynomial matrix. The notions of finite and infinite Jordan pairs were found originally in [5] and are based on the notions of finite and infinite Jordan chains [4],[5]. Our analysis is based on the theory presented in recent papers [3],[9],[10], where simple and efficient methods of finding finite and infinite Jordan chains—and as a consequence Jordan pairs—using the notions of finite and infinite elementary divisors, are given. Specifically we prove the following:

**Proposition 1** Let  $A(s) \in \mathbb{R}^{r \times r}[s]$  be a polynomial matrix. Let also the finite Jordan pair  $C_f \in \mathbb{R}^{r \times n}$ ,  $J_f \in \mathbb{R}^{n \times n}$ , and the matrix  $B_f \in \mathbb{R}^{n \times r}$  with  $n = \deg |A(s)|$ , such that  $H_{spr}(s) = C_f[sI_n - J_f]^{-1} B_f$  where  $H_{spr}(s)$  is the strictly proper part of  $A(s)^{-1}$ . Then the triple  $(C_f, J_f, B_f)$  is a minimal realization for  $H_{spr}(s)$ .  $\square$

**Proposition 2** Let  $A(s) \in \mathbb{R}^{r \times r}[s]$  be a polynomial matrix. Let also the infinite Jordan pair  $(C_\infty, J_\infty)$  of  $A(s)$ , with  $C_\infty \in \mathbb{R}^{r \times \mu}$ ,  $J_\infty \in \mathbb{R}^{\mu \times \mu}$ , where  $\mu$  is given by  $\mu = (r-1)q_1 - \sum_{i=2}^k q_i + \sum_{j=k+1}^r \hat{q}_j$  where  $q_i$ ,

$i = 1, \dots, k$  and  $\hat{q}_j$ ,  $j = k+1, \dots, r$  denote the orders of the poles and zeros at  $s = \infty$  of  $A(s)$  respectively and  $B_\infty \in \mathbb{R}^{\mu \times r}$ . Let also  $A(s)^{-1} = H_{spr}(s) + H_{pol}(s)$ . Then (i) the triple of matrices  $(C_\infty, J_\infty, B_\infty)$  is a realization for the polynomial part  $H_{pol}(s)$  of  $A(s)^{-1}$ . (ii) From  $(C_\infty, J_\infty, B_\infty)$  we can find a triple of matrices  $(\tilde{C}_\infty, \tilde{J}_\infty, \tilde{B}_\infty)$ , with  $\tilde{C}_\infty \in \mathbb{R}^{r \times \tilde{\mu}}$ ,  $\tilde{J}_\infty \in \mathbb{R}^{\tilde{\mu} \times \tilde{\mu}}$ ,  $\tilde{B}_\infty \in \mathbb{R}^{\tilde{\mu} \times r}$  where  $\tilde{\mu}$  is given by  $\tilde{\mu} = (r-k) + \sum_{j=k+1}^r \hat{q}_j$

Clearly  $\tilde{\mu} \leq \mu$  and the triple  $(\tilde{C}_\infty, \tilde{J}_\infty, \tilde{B}_\infty)$  constitutes a minimal realization of the polynomial part  $H_{pol}(s)$  of  $A(s)^{-1}$ , i.e.  $H_{pol}(s) = \tilde{C}_\infty[s\tilde{J}_\infty - I_{\tilde{\mu}}]^{-1} \tilde{B}_\infty$  and  $\mu = \delta_M[1/w A(1/w)]$   $\square$

The proposed method can be applied to the so-called realization theory of transfer function matrices of Linear Multivariable Systems [6], i.e. physical systems of the form  $(\Sigma): A(\rho)\beta(t) = B(\rho)u(t)$   $y(t) = C(\rho)\beta(t)$ , where  $\rho := d/dt$  is the differential operator,  $A(\rho)$ ,  $B(\rho)$ ,  $C(\rho)$  are polynomial matrices and  $\beta(t)$ ,  $y(t)$ ,  $u(t)$  are respectively the pseudostate, the output and the input vectors of the system  $(\Sigma)$ . The transfer function matrix of  $(\Sigma)$  is (in frequency-domain):  $G(s) = C(s)A(s)^{-1}B(s)$  which is a

rational matrix (not necessarily proper) in general. It would be interesting to find certain singular systems in generalized state-space form [1], i.e. physical systems of the form  $(\Sigma_1) : E\dot{x}(t) = Ax(t) + Bu(t)$ ,

$y(t) = Cx(t)$  — where  $E, A, B, C$  are constant matrices with appropriate dimensions and  $x(t), y(t), u(t)$  are respectively the generalized state, the output and the input vectors of the system  $(\Sigma_1)$  — which give rise to the transfer function matrix  $G(s)$ . In other words the transfer function matrix of system  $(\Sigma_1)$  which is given by:  $G_1(s) = C[sE - A]^{-1}B$  satisfies the following condition:  $G_1(s) = C[sE - A]^{-1}B = C(s)A(s)^{-1}B(s) = G(s)$ .

**Definition 1** [2] Assume that  $G_1(s) \in \mathbb{R}^{r \times r}(s)$  is a rational matrix. If there exists a quadruple of matrices  $(E, A, B, C)$  such that:  $G_1(s) = C[sE - A]^{-1}B$  — where  $E, A \in \mathbb{R}^{\hat{n} \times \hat{n}}$ ,  $B \in \mathbb{R}^{\hat{n} \times r}$ ,  $C \in \mathbb{R}^{r \times \hat{n}}$  are constant matrices with  $\hat{n} \in \mathbb{N} - \{0\}$  —, then the generalized state-space system described by  $(\Sigma_1)$  will be called a singular system realization of  $G_1(s)$ , or simply a realization of  $G_1(s)$ . Furthermore the system  $(\Sigma_1)$  is called a minimal realization of  $G_1(s)$  iff any other realization of  $G_1(s)$  has order greater than  $\hat{n}$ , or equivalently iff the generalized state-space system  $(\Sigma_1)$  has the least number of generalized states  $x(t)$ .

Any rational matrix  $G(s)$  (not necessarily proper) may be represented as the sum of its strictly proper part  $H_{spr}(s)$  and its polynomial part  $H_{pol}(s)$ , i.e.  $G(s) = H_{spr}(s) + H_{pol}(s)$ . We know that the

inverse of a polynomial matrix  $F(s) \in \mathbb{R}^{r \times r}[s]$  is a rational matrix in general. If we now consider the case where  $F(s)^{-1} = G(s)$  then the proposed method finds a minimal realization — as this defined in definition 1 — of a transfer function matrix  $G(s)$  of a system  $(\Sigma)$  which has the property its inverse to be a polynomial matrix. In other words we can find a quadruple of matrices  $[E, A, B, C]$  which have the following properties: (i) give rise to the generalized state-space system  $(\Sigma_1)$ , and (ii)  $G(s) = C[sE - A]^{-1}B$ .

To be more precise let a system  $(\Sigma)$  which give rise to a transfer function matrix  $G(s) = C(s)A(s)^{-1}B(s) \in \mathbb{R}^{r \times r}(s)$ , and assume that  $G(s)$  has the following property  $G(s)^{-1} = F(s) \in \mathbb{R}^{r \times r}[s]$ . Now proposition 1 states that we can find a triple of matrices  $C_f \in \mathbb{R}^{r \times n}$ ,  $J_f \in \mathbb{R}^{n \times n}$ ,  $B_f \in \mathbb{R}^{n \times r}$  with  $n = \deg |F(s)|$ , such that:  $H_{spr}(s) = C_f[sI_n - J_f]^{-1}B_f$ , where  $H_{spr}(s)$  is the strictly proper part of  $G(s) = F(s)^{-1}$  and the triple  $(C_f, J_f, B_f)$  is a minimal realization of  $H_{spr}(s)$ . Also proposition 2 states that we can

find a triple of matrices  $(\tilde{C}_\infty, \tilde{J}_\infty, \tilde{B}_\infty)$ , with  $\tilde{C}_\infty \in \mathbb{R}^{r \times \tilde{\mu}}$ ,  $\tilde{J}_\infty \in \mathbb{R}^{\tilde{\mu} \times \tilde{\mu}}$ ,  $\tilde{B}_\infty \in \mathbb{R}^{\tilde{\mu} \times r}$  with  $\tilde{\mu} = (r-k) + \sum_{j=k+1}^r \hat{q}_j$  such that  $H_{pol}(s) = \tilde{C}_\infty[s\tilde{J}_\infty - I_{\tilde{\mu}}]^{-1}\tilde{B}_\infty$ , where  $H_{pol}(s)$  is the polynomial part of  $G(s) = F(s)^{-1}$  and the

triple  $(\tilde{C}_\infty, \tilde{J}_\infty, \tilde{B}_\infty)$  is a minimal realization of  $H_{pol}(s)$ . Let now define  $E := \begin{bmatrix} I_n & 0 \\ 0 & J_\infty \end{bmatrix}$ ,  $A := \begin{bmatrix} J_f & 0 \\ 0 & I_{\tilde{\mu}} \end{bmatrix}$ ,  $B := \begin{bmatrix} B_f \\ \tilde{B}_\infty \end{bmatrix}$ ,  $C := \begin{bmatrix} C_f & \tilde{C}_\infty \end{bmatrix}$ . We can now define the generalized state-space system  $(\Sigma_1)$  with

$E, A, B, C$  as above. It is easy to verify that:  $G(s) = H_{spr}(s) + H_{pol}(s) = C[sE - A]^{-1}B$

Hence the system  $(\Sigma_1)$  determined by the matrices  $[E, A, B, C]$  is a realization of  $G(s)$ . Furthermore we can prove easily that the above realization is also a minimal one.

**Definition 2** The order  $\hat{n}$  of the minimal realization of the transfer function matrix  $G(s)$  of  $(\Sigma)$  is called the minimum generalized order of  $G(s)$ . Furthermore  $\hat{n}$  is the dimension of the generalized state-space system  $(\Sigma_1)$  and is equal to :  $\hat{n} = n + \tilde{\mu} = \deg |F(s)| + (r-k) + \sum_{j=k+1}^r \hat{q}_j$ , where  $\hat{q}_j$ ,  $j = k+1, \dots, r$

denote the orders of the zeros at  $s = \infty$  of the polynomial matrix  $F(s)$  which can be found using the Smith-McMillan form at  $s = \infty$  [8]. We can now state the following.

**Theorem 1** Let a linear multivariable system  $(\Sigma)$  which give rise to a rational transfer function matrix  $G(s) \in \mathbb{R}^{r \times r}(s)$  and has the property to have a polynomial inverse  $F(s) \in \mathbb{R}^{r \times r}[s]$ . Then we can find a generalized state-space system of the form  $(\Sigma_1)$  and minimum generalized order  $\hat{n}$ , such that the system  $(\Sigma_1)$  to be a minimal realization of the rational matrix  $G(s)$  (according to definition 1). Furthermore since the two systems  $(\Sigma)$  and  $(\Sigma_1)$  give rise to the same transfer function matrix  $G(s)$  they have the same sets of finite and infinite transmission poles and zeros ([6],[7]).  $\square$

We remark here that the problem of transforming a linear multivariable system  $(\Sigma)$  to a generalized state space system  $(\Sigma_1)$  is called linearization and has been considered by many researchers.

In our paper we study a special case of linearization ; that is linearization for the class of transfer function matrices  $G(s)$  of systems  $(\Sigma)$  with the property of having a polynomial inverse, i.e.:

$$\mathcal{V} = \{ G(s) \in \mathbb{R}^{r \times r}(s) / G(s)^{-1} \in \mathbb{R}^{r \times r}[s] \}$$

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